

Delay-dependent LMI condition to the stability of a category of uncertain discrete-time systems exerting quantization/overflow nonlinearities and variable time-lags

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Abstract

This paper deals with the stability of nonlinear uncertain discrete-time systems with variable time-lags. The system under assumption involves norm-bounded parameter uncertainties, quantization/overflow nonlinearities and variable time-lags. A Lyapunov function based delay-dependent sufficient stability condition is derived by using a more relaxed finite-sum inequality. As a special case, a stability condition for constant delay is also derived in this paper. Finally, numerical examples are given to show the efficacy of the proposed conditions.

Keywords

Finite wordlength nonlinearity, Linear matrix inequality, Lyapunov method, Norm-bounded parameter uncertainty, Variable time-lag.

1. Introduction

While implementing the category of discrete-time systems via specific binary hardware, there is a chance of occurrence of nonlinearities for example quantization and/or overflow in the designed system. Due to such nonlinear effects, the unpredictability in the system occurs in the form of limit cycles or unwanted oscillations that may tend the system to be unstable. So, it is a key issue to find the range of system parameters under which the designed system is not affected by limit cycles. During the past, several studies available on the composite effects of quantization and overflow [1-6] for discrete realizable systems.

Another source of instability in the designed system is parameter uncertainty which arises due to several reasons like finite resolution of measuring equipment, variation in system parameters, modeling error or some ignored factor. The effects of parameter uncertainties are studied extensively in existing literature [3-18].

Besides nonlinearities and parameter uncertainties, another source of instability in discrete-time systems is time-lag. Time-lag occurs due to finite computational time or transportation lag required for the transmission of information among the various parts of the system. The generated time-lag in processing of information may be constant or varying with time. Available stability conditions in literature are classified into two different categories in which one is delay-independent [3, 10] and other is delay-dependent [4, 5, 6, 8, 13, 15, 17-32]. Usually, delay-dependent conditions bring less

conservative results because these conditions utilize the information of the size of the lag whereas delay-independent conditions have not included the information of the size of the lag. For expecting less conservative results, selection of appropriate Lyapunov function and application of tighter bounding techniques in the sum and cross terms of the forward difference of the Lyapunov function are the key steps. Various tighter bounding techniques from the existing literature are free-weighting matrix method [20, 33], Jensen inequality [21], Reciprocal convex method [6, 24], Wirtinger-based inequality [6, 17, 34] and Abel lemma-based inequality [35].

Practical engineering systems such as network control systems [14], discrete-time Markovian jump systems [22], discrete-time neural networks [2], Sensor networks [36] etc. are frequently suffered by the presence of above discussed instability factors. The problem of stability investigation of a category of discrete-time systems exerting quantization and/or overflow nonlinearities, parameter uncertainties and variable time-lags is more interesting and realistic in nature. Many results [3-6, 10, 17] have previously reported for the global asymptotic stability of these systems, yet there is a scope to achieve improved results over the previous results.

The key objective of the paper is to derive less conservative delay-dependent conditions for the global asymptotic stability of a category of uncertain discrete-time systems employing quantization and/or overflow

nonlinearities and variable time-lags. The main contributions of this work are as follows:

1. A new global asymptotic stability criterion for uncertain discrete-time systems having quantization and/or overflow nonlinearities and variable time-lag is derived. The criterion utilizes a more relaxed technique [26, 28] which may reduce the conservatism and simplify the system analysis/synthesis process. As a special case of our main result, a stability condition for the category of uncertain discrete-time systems with constant time-lag and quantization/overflow nonlinearities is established.
2. The proposed conditions are in the setting of linear matrix inequalities (LMIs), and hence, one can easily test the conditions by using well-known LMI solvers [37, 39].
3. A comparison of the proposed results with the existing results [4-6] is given.

The remaining paper is outlined as follows. In Section 2, the system description and some useful lemmas are provided while in Section 3, the main results of the paper are derived. In Section 4, two numerical examples to demonstrate the effectiveness of the main results are given. Section 5 provides the conclusion of the paper and in Section 6, the future scope of the presented work is highlighted.

Notations: $\mathbf{0}$ is the null matrix or null vector; \mathbf{I} is the identity matrix; $\mathbf{R}^{\alpha \times \beta}$ is the set of $\alpha \times \beta$ real matrices; \mathbf{R}^α is the set of $\alpha \times 1$ real vectors; \mathbf{M}^T denotes the transpose of matrix \mathbf{M} ; $\mathbf{M} > \mathbf{0}$ ($\mathbf{M} < \mathbf{0}$) means that \mathbf{M} is positive (negative) definite symmetric matrix; maximum and minimum eigen values of a matrix \mathbf{M} are $\lambda_{\max}(\mathbf{M})$ and $\lambda_{\min}(\mathbf{M})$, respectively; $\text{diag}(\mathbf{z}_1, \mathbf{z}_2, \dots, \mathbf{z}_g)$ is a diagonal matrix with diagonal entries $\mathbf{z}_1, \mathbf{z}_2, \dots, \mathbf{z}_g$; the symmetric entries of a symmetric matrix is inferred by $*$.

2 System Description

The system under inspection is as

$$\begin{aligned} \chi(u+1) &= \mathbf{O}\{\mathbf{Q}(\gamma(u))\} = \boldsymbol{\lambda}(\gamma(u)) \\ &= [\lambda_1(\gamma_1(u)) \ \lambda_2(\gamma_2(u)) \ \dots \ \lambda_n(\gamma_n(u))]^T, \end{aligned} \quad (1a)$$

$$\gamma(u) = (\mathbf{A} + \Delta\mathbf{A})\chi(u) + (\mathbf{A}_d + \Delta\mathbf{A}_d)\chi(u-d(u)) \quad (1b)$$

$$\begin{aligned} &= [\gamma_1(u) \ \gamma_2(u) \ \dots \ \gamma_n(u)]^T, \\ \chi(u) &= \boldsymbol{\varphi}(u), \forall u \in [-h_2, 0] \end{aligned} \quad (1c)$$

where $\chi(u) \in \mathbf{R}^n$ is the state vector; $\mathbf{A}, \mathbf{A}_d \in \mathbf{R}^{n \times n}$ are the known constant matrices; The unknown matrices

$\Delta\mathbf{A}, \Delta\mathbf{A}_d \in \mathbf{R}^{n \times n}$ denoting parametric uncertainties in the state matrices; $\mathbf{Q}(\cdot)$ and $\mathbf{O}(\cdot)$ signify quantization and overflow nonlinearities, respectively; $\boldsymbol{\lambda}(\cdot)$ denotes composite nonlinear effects including both quantization and overflow. Here, $\mathbf{Q}(\cdot)$ being either roundoff or magnitude truncation and $\boldsymbol{\lambda}(\cdot)$ is confined to the sector $[k_o, k_q]$, that is,

$$\lambda_i(0) = 0, \quad k_o \gamma_i^2(u) \leq \lambda_i(\gamma_i(u)) \gamma_i(u) \leq k_q \gamma_i^2(u), \quad i=1, 2, \dots, n \quad (2a)$$

where

$$\begin{aligned} k_q &= \begin{cases} 1, & \text{for magnitude truncation,} \\ 2, & \text{for roundoff,} \end{cases} \\ k_o &= \begin{cases} 0, & \text{for saturation or zeroing,} \\ -\frac{1}{3}, & \text{for triangular,} \\ -1, & \text{for 2's complement,} \end{cases} \end{aligned} \quad (2b)$$

and $d(u)$ is the variable time-lag that satisfy

$$1 \leq h_1 \leq d(u) \leq h_2 \quad (3)$$

where h_1 and h_2 are the lower and upper admissible time-lag limits, respectively. $\Delta\mathbf{A}, \Delta\mathbf{A}_d \in \mathbf{R}^{n \times n}$ are considered as

$$\Delta\mathbf{A} = \mathbf{H}_0 \mathbf{F}_0 \mathbf{E}_0, \quad (4a)$$

$$\Delta\mathbf{A}_d = \mathbf{H}_1 \mathbf{F}_1 \mathbf{E}_1 \quad (4b)$$

where $\mathbf{H}_i \in \mathbf{R}^{n \times p_i}$, $\mathbf{E}_i \in \mathbf{R}^{q_i \times n}$ ($i = 0, 1$) are constant matrices (known) and $\mathbf{F}_i \in \mathbf{R}^{p_i \times q_i}$ ($i = 0, 1$) is a matrix (unknown) which fulfills

$$\mathbf{F}_i^T \mathbf{F}_i \leq \mathbf{I}, \quad i = 0, 1. \quad (4c)$$

Lemma 1 [26, 28] For a matrix $\mathbf{R} = \mathbf{R}^T > \mathbf{0}$, any matrices \mathbf{L}_1 and \mathbf{L}_2 of appropriate dimensions, integers a_1 and a_2 satisfying $a_1 < a_2$ such that (5) holds

$$\begin{aligned} & - \sum_{s=u-a_2}^{u-a_1-1} \eta^T(s) \mathbf{R} \eta(s) \leq (a_2 - a_1) \begin{bmatrix} \mathbf{k}_1 \\ \mathbf{k}_2 \end{bmatrix}^T \begin{bmatrix} \mathbf{\Omega}_1 & \mathbf{0} \\ \mathbf{0} & \mathbf{\Omega}_2 \end{bmatrix} \begin{bmatrix} \mathbf{k}_1 \\ \mathbf{k}_2 \end{bmatrix} \\ & - 2 \begin{bmatrix} \mathbf{k}_1 \\ \mathbf{k}_2 \end{bmatrix}^T \begin{bmatrix} \mathbf{L}_1 & \mathbf{0} \\ \mathbf{0} & \mathbf{L}_2 \end{bmatrix} \begin{bmatrix} \mathbf{k}_1 \\ \mathbf{k}_2 \end{bmatrix} \end{aligned} \quad (5)$$

where

$$\Gamma(s) = \chi(s+1) - \chi(s),$$

$$k_1 = \chi(u - a_1) - \chi(u - a_2),$$

$$k_2 = \chi(u - a_1) + \chi(u - a_2) - \sum_{i=u-a_2}^{u-a_1} \frac{2\chi(i)}{a_2 - a_1 + 1},$$

$$\Omega_1 = L_1 R^{-1} L_1^T \text{ and } \Omega_2 = L_2 (3R)^{-1} L_2^T.$$

Lemma 2 [16, 37, 38] Let Σ, A, F and M be real matrices of appropriate sizes with M fulfilling $M = M^T$, then

$$M + \Sigma F A + A^T F^T \Sigma^T < 0 \tag{6}$$

for all $F^T F \leq I$, if and only if there exists a scalar $\varepsilon > 0$ satisfying

$$M + \varepsilon^{-1} \Sigma \Sigma^T + \varepsilon A^T A < 0. \tag{7}$$

Remark 1 In [26], Lemma 1 has been used for stability analysis of discrete-time systems having variable time-lag but not having quantization/overflow nonlinearities and parameter uncertainties.

Now, we are presenting the stability conditions of the paper.

3 Main Result

Theorem 1. For known integers h_1 and h_2 , the system (1)-(4) is globally asymptotically stable (GAS) if there

exist matrices $0 < P = \begin{bmatrix} P_1 & P_2 & P_3 & P_4 \\ * & P_5 & P_6 & P_7 \\ * & * & P_8 & P_9 \\ * & * & * & P_{10} \end{bmatrix} = P^T \in \mathbb{R}^{4n \times 4n}$,

$0 < M_i = M_i^T (i = 1, 2, 3) \in \mathbb{R}^{n \times n}$, $0 < N_i = N_i^T (i = 1, 2) \in \mathbb{R}^{n \times n}$, any matrices $L_i (i = 1, 2, 3, \dots, 6) \in \mathbb{R}^{n \times n}$, $0 < G = G^T = \text{diag}(g_1, g_2, \dots, g_n)$ and two positive scalars $\varepsilon_0, \varepsilon_1$ such that LMIs (8a) and (8b) hold simultaneously:

$$\Psi(d(u) = h_1) < 0, \tag{8a}$$

$$\Psi(d(u) = h_2) < 0 \tag{8b}$$

where

$$\Psi(d(u)) = \begin{bmatrix} Q_{11} + \varepsilon_0 E_0^T E_0 & Q_{12} & -P_7 + P_4 & 0 & Q_{15} & Q_{16} \\ * & Q_{22} & Q_{23} & P_7 & Q_{25} & Q_{26} \\ * & * & Q_{33} + \varepsilon_1 E_1^T E_1 & Q_{34} & Q_{35} & Q_{36} \\ * & * & * & Q_{44} & -P_7^T (h_1 + 1) & Q_{46} \\ * & * & * & * & -4h_1(L_2 + L_2^T) & 0 \\ * & * & * & * & * & Q_{66} \\ * & * & * & * & * & * \\ * & * & * & * & * & * \\ * & * & * & * & * & * \\ * & * & * & * & * & * \\ * & * & * & * & * & * \\ * & * & * & * & * & * \\ * & * & * & * & * & * \\ * & * & * & * & * & * \\ * & * & * & * & * & * \\ * & * & * & * & * & * \\ * & * & * & * & * & * \end{bmatrix}$$

$$\begin{bmatrix} Q_{17} & Q_{18} & h_1 L_1 & h_1 L_2 & 0 \\ Q_{27} & -P_2^T & -h_1 L_1 & h_1 L_2 & Q_{211} L_3 \\ Q_{37} & -P_3^T + k_q A_d^T G & 0 & 0 & -Q_{211} L_3 \\ Q_{47} & -P_4^T & 0 & 0 & 0 \\ 0 & P_2^T (h_1 + 1) & 0 & -2h_1 L_2 & 0 \\ 0 & Q_{68} & 0 & 0 & 0 \\ Q_{77} & Q_{78} & 0 & 0 & 0 \\ * & Q_{88} & 0 & 0 & 0 \\ * & * & -N_1 & 0 & 0 \\ * & * & * & -3N_1 & 0 \\ * & * & * & * & -N_2 \\ * & * & * & * & * \\ * & * & * & * & * \\ * & * & * & * & * \\ * & * & * & * & * \\ * & * & * & * & * \end{bmatrix}$$

$$\begin{bmatrix} 0 & 0 & 0 & Q_{115} & 0 & 0 \\ 0 & Q_{211} L_4 & 0 & 0 & 0 & 0 \\ Q_{312} L_5 & Q_{211} L_4 & Q_{312} L_6 & Q_{315} & 0 & 0 \\ -Q_{312} L_5 & 0 & Q_{312} L_6 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & -2Q_{211} L_4 & 0 & 0 & 0 & 0 \\ 0 & 0 & -2Q_{312} L_6 & 0 & 0 & 0 \\ 0 & 0 & 0 & \sqrt{-k_q} / 2 G & k_q G H_0 & k_q G H_1 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ -N_2 & 0 & 0 & 0 & 0 & 0 \\ * & -3N_2 & 0 & 0 & 0 & 0 \\ * & * & -3N_2 & 0 & 0 & 0 \\ * & * & * & -k_q G & -k_q \sqrt{-2k_q} G H_0 & -k_q \sqrt{-2k_q} G H_1 \\ * & * & * & * & -\varepsilon_0 I & 0 \\ * & * & * & * & * & -\varepsilon_1 I \end{bmatrix}, \tag{9}$$

$$\mathbf{Q}_{11} = -\mathbf{P}_5 - \mathbf{P}_1 + \mathbf{P}_2^T + \mathbf{P}_2 + \mathbf{M}_1 + h_1^2 \mathbf{N}_1 + h_{12}^2 \mathbf{N}_2 - h_1(\mathbf{L}_1^T + \mathbf{L}_1 + \mathbf{L}_2^T + \mathbf{L}_2) + (h_{12} + 1)\mathbf{M}_3, \quad (10)$$

$$\mathbf{Q}_{12} = \mathbf{P}_3 - \mathbf{P}_6 + h_1(\mathbf{L}_1^T + \mathbf{L}_1 - \mathbf{L}_2^T - \mathbf{L}_2), \quad (11)$$

$$\mathbf{Q}_{15} = (h_1 + 1)(\mathbf{P}_5 - \mathbf{P}_2) + 2h_1(\mathbf{L}_2^T + \mathbf{L}_2), \quad (12)$$

$$\mathbf{Q}_{16} = (d(u) - h_1 + 1)(\mathbf{P}_6 - \mathbf{P}_3), \quad (13)$$

$$\mathbf{Q}_{17} = (h_2 - d(u) + 1)(\mathbf{P}_7 - \mathbf{P}_4), \quad (14)$$

$$\mathbf{Q}_{18} = -(h_1^2 \mathbf{N}_1 + h_{12}^2 \mathbf{N}_2) + k_q \mathbf{A}^T \mathbf{G}, \quad (15)$$

$$\mathbf{Q}_{115} = -k_q \sqrt{-2k_0} \mathbf{A}^T \mathbf{G}, \quad (16)$$

$$\mathbf{Q}_{22} = \mathbf{P}_5 - \mathbf{P}_8 + \mathbf{M}_2 - \mathbf{M}_1 - h_1(\mathbf{L}_1 + \mathbf{L}_2 + \mathbf{L}_1^T + \mathbf{L}_2^T) - h_{12}(\mathbf{L}_3 + \mathbf{L}_4 + \mathbf{L}_3^T + \mathbf{L}_4^T), \quad (17)$$

$$\mathbf{Q}_{23} = \mathbf{P}_6 - \mathbf{P}_9 + h_{12}(\mathbf{L}_3^T + \mathbf{L}_3 - \mathbf{L}_4^T - \mathbf{L}_4), \quad (18)$$

$$\mathbf{Q}_{25} = (-\mathbf{P}_5 + \mathbf{P}_6^T)(h_1 + 1) + 2h_1(\mathbf{L}_2^T + \mathbf{L}_2), \quad (19)$$

$$\mathbf{Q}_{26} = (d(u) - h_1 + 1)(\mathbf{P}_8 - \mathbf{P}_6) + 2h_{12}(\mathbf{L}_4^T + \mathbf{L}_4), \quad (20)$$

$$\mathbf{Q}_{27} = (\mathbf{P}_9 - \mathbf{P}_7)(h_2 - d(u) + 1), \quad (21)$$

$$\mathbf{Q}_{211} = \sqrt{h_{12}} \sqrt{d(u) - h_1}, \quad (22)$$

$$\mathbf{Q}_{312} = \sqrt{h_{12}} \sqrt{h_2 - d(u)}, \quad (23)$$

$$\mathbf{Q}_{315} = -k_q \sqrt{-2k_0} \mathbf{A}_d^T \mathbf{G}, \quad (24)$$

$$\mathbf{Q}_{33} = \mathbf{P}_8 - \mathbf{P}_{10} - \mathbf{M}_3 - h_{12}(\mathbf{L}_3 + \mathbf{L}_4 + \mathbf{L}_5 + \mathbf{L}_6 + \mathbf{L}_3^T + \mathbf{L}_4^T + \mathbf{L}_5^T + \mathbf{L}_6^T), \quad (25)$$

$$\mathbf{Q}_{34} = \mathbf{P}_9 + h_{12}(\mathbf{L}_5^T + \mathbf{L}_5 - \mathbf{L}_6^T - \mathbf{L}_6), \quad (26)$$

$$\mathbf{Q}_{35} = (h_1 + 1)(\mathbf{P}_7^T - \mathbf{P}_6^T), \quad (27)$$

$$\mathbf{Q}_{36} = (d(u) - h_1 + 1)(\mathbf{P}_9^T - \mathbf{P}_8^T) + 2h_{12}(\mathbf{L}_4^T + \mathbf{L}_4), \quad (28)$$

$$\mathbf{Q}_{37} = (h_2 - d(u) + 1)(\mathbf{P}_{10} - \mathbf{P}_9) + 2h_{12}(\mathbf{L}_6^T + \mathbf{L}_6), \quad (29)$$

$$\mathbf{Q}_{44} = \mathbf{P}_{10} - \mathbf{M}_2 - h_{12}(\mathbf{L}_5^T + \mathbf{L}_5 + \mathbf{L}_6^T + \mathbf{L}_6), \quad (30)$$

$$\mathbf{Q}_{46} = -\mathbf{P}_9^T (d(u) - h_1 + 1), \quad (31)$$

$$\mathbf{Q}_{47} = -\mathbf{P}_{10} (h_2 - d(u) + 1) + 2h_{12}(\mathbf{L}_6^T + \mathbf{L}_6), \quad (32)$$

$$\mathbf{Q}_{66} = -4h_{12}(\mathbf{L}_4 + \mathbf{L}_4^T), \quad (33)$$

$$\mathbf{Q}_{68} = \mathbf{P}_3^T (d(u) - h_1 + 1), \quad (34)$$

$$\mathbf{Q}_{77} = -4h_{12}(\mathbf{L}_6 + \mathbf{L}_6^T), \quad (35)$$

$$\mathbf{Q}_{78} = \mathbf{P}_4^T (h_2 - d(u) + 1), \quad (36)$$

$$h_{12} = h_2 - h_1, \quad (37)$$

$$\mathbf{Q}_{88} = \mathbf{P}_1 + h_1^2 \mathbf{N}_1 + h_{12}^2 \mathbf{N}_2 + [(k_0 / 2k_q) - 2]\mathbf{G}. \quad (38)$$

Proof. Consider a quadratic Lyapunov function

$$\begin{aligned} V(\boldsymbol{\chi}(u)) = & \boldsymbol{\eta}^T(u) \mathbf{P} \boldsymbol{\eta}(u) + \sum_{s=u-h_1}^{u-1} \boldsymbol{\chi}^T(s) \mathbf{M}_1 \boldsymbol{\chi}(s) + \sum_{s=u-h_2}^{u-h_1-1} \boldsymbol{\chi}^T(s) \mathbf{M}_2 \boldsymbol{\chi}(s) \\ & + \sum_{\theta=-h_2}^{-h_1} \sum_{s=u+\theta}^{u-1} \boldsymbol{\chi}^T(s) \mathbf{M}_3 \boldsymbol{\chi}(s) + h_1 \sum_{\theta=-h_1}^{-1} \sum_{s=u+\theta}^{u-1} \boldsymbol{\Gamma}^T(s) \mathbf{N}_1 \boldsymbol{\Gamma}(s) \\ & + h_{12} \sum_{\theta=-h_2}^{-h_1-1} \sum_{s=u+\theta}^{u-1} \boldsymbol{\Gamma}^T(s) \mathbf{N}_2 \boldsymbol{\Gamma}(s) \end{aligned} \quad (39)$$

where

$$\boldsymbol{\eta}^T(u) = [\boldsymbol{\chi}^T(u) \sum_{s=u-h_1}^{u-1} \boldsymbol{\chi}^T(s) \sum_{s=u-d(u)}^{u-h_1-1} \boldsymbol{\chi}^T(s) \sum_{s=u-h_2}^{u-d(u)-1} \boldsymbol{\chi}^T(s)] \quad (40)$$

and

$$\boldsymbol{\Gamma}(u) = \boldsymbol{\chi}(u+1) - \boldsymbol{\chi}(u) = \boldsymbol{\lambda}(\boldsymbol{\chi}(u)) - \boldsymbol{\chi}(u). \quad (41)$$

Application of (39) to (1) yields

$$\begin{aligned} \Delta V(\boldsymbol{\chi}(u)) = & V(\boldsymbol{\chi}(u+1)) - V(\boldsymbol{\chi}(u)) \\ = & \boldsymbol{\eta}^T(u+1) \mathbf{P} \boldsymbol{\eta}(u+1) - \boldsymbol{\eta}^T(u) \mathbf{P} \boldsymbol{\eta}(u) \\ & + \boldsymbol{\chi}^T(u) \mathbf{M}_1 \boldsymbol{\chi}(u) - \boldsymbol{\chi}^T(u-h_1) (\mathbf{M}_1 - \mathbf{M}_2) \boldsymbol{\chi}(u-h_1) \\ & - \boldsymbol{\chi}^T(u-h_2) \mathbf{M}_2 \boldsymbol{\chi}(u-h_2) \\ & + \boldsymbol{\chi}^T(u) \mathbf{M}_3 \boldsymbol{\chi}(u) + h_{12} \boldsymbol{\chi}^T(u) \mathbf{M}_3 \boldsymbol{\chi}(u) \\ & - \sum_{s=u-h_2}^{u-h_1} \boldsymbol{\chi}^T(s) \mathbf{M}_3 \boldsymbol{\chi}(s) \\ & + \boldsymbol{\Gamma}^T(u) [h_1^2 \mathbf{N}_1 + h_{12}^2 \mathbf{N}_2] \boldsymbol{\Gamma}(u) \\ & - h_1 \sum_{s=u-h_1}^{u-1} \boldsymbol{\Gamma}^T(s) \mathbf{N}_1 \boldsymbol{\Gamma}(s) \\ & - h_{12} \sum_{s=u-h_2}^{u-h_1-1} \boldsymbol{\Gamma}^T(s) \mathbf{N}_2 \boldsymbol{\Gamma}(s). \end{aligned} \quad (42)$$

Now, (42) can be further rearranged as

$$\begin{aligned} \Delta V(\boldsymbol{\chi}(u)) = & \boldsymbol{\Xi}^T(u) \boldsymbol{\Theta}(d(u)) \boldsymbol{\Xi}(u) + \boldsymbol{\chi}^T(u) \mathbf{M}_1 \boldsymbol{\chi}(u) \\ & - \boldsymbol{\chi}^T(u-h_1) (\mathbf{M}_1 - \mathbf{M}_2) \boldsymbol{\chi}(u-h_1) \\ & - \boldsymbol{\chi}^T(u-h_2) \mathbf{M}_2 \boldsymbol{\chi}(u-h_2) + \boldsymbol{\chi}^T(u) \mathbf{M}_3 \boldsymbol{\chi}(u) \\ & + h_{12} \boldsymbol{\chi}^T(u) \mathbf{M}_3 \boldsymbol{\chi}(u) - \sum_{s=u-h_2}^{u-h_1} \boldsymbol{\chi}^T(s) \mathbf{M}_3 \boldsymbol{\chi}(s) \\ & + \boldsymbol{\Gamma}^T(u) [h_1^2 \mathbf{N}_1 + h_{12}^2 \mathbf{N}_2] \boldsymbol{\Gamma}(u) \\ & - h_1 \sum_{s=u-h_1}^{u-1} \boldsymbol{\Gamma}^T(s) \mathbf{N}_1 \boldsymbol{\Gamma}(s) - h_{12} \sum_{s=u-h_2}^{u-h_1-1} \boldsymbol{\Gamma}^T(s) \mathbf{N}_2 \boldsymbol{\Gamma}(s) \end{aligned} \quad (43)$$

where

$$\begin{aligned} \boldsymbol{\Xi}^T(u) = & \left[\boldsymbol{\chi}^T(u) \boldsymbol{\chi}^T(u-h_1) \boldsymbol{\chi}^T(u-d(u)) \boldsymbol{\chi}^T(u-h_2) \frac{1}{h_1+1} \sum_{s=u-h_1}^u \boldsymbol{\chi}^T(s) \right. \\ & \left. \frac{1}{d(u)-h_1+1} \sum_{s=u-d(u)}^{u-h_1} \boldsymbol{\chi}^T(s) \frac{1}{h_2-d(u)+1} \sum_{s=u-h_2}^{u-d(u)} \boldsymbol{\chi}^T(s) \boldsymbol{\lambda}^T(\boldsymbol{\gamma}(u)) \right], \end{aligned} \quad (44)$$

Example 1. Choose the system (1) - (4) with

$$\mathbf{A} = \begin{bmatrix} 0.8 & 0 \\ 0.05 & 0.9 \end{bmatrix}, \mathbf{A}_d = \begin{bmatrix} -0.1 & 0 \\ -0.2 & -0.1 \end{bmatrix} \quad (68a)$$

$$\mathbf{H}_0 = \mathbf{H}_1 = \begin{bmatrix} 0 \\ 0.1 \end{bmatrix}, \mathbf{E}_0 = [0.01 \ 0], \mathbf{E}_1 = [0 \ 0.01] \quad (68b)$$

and the nonlinearities belongs to the sector $[k_o, k_q] = [0 \ 1]$, which comprises zeroing, saturation, magnitude truncation, a combination of zeroing and magnitude truncation, a combination of saturation and magnitude truncation, and so on. In [4-6], this example has been considered. Table 1 shows the upper lag limit (h_2) for several given lower lag limits (h_1). In Table 1, while showing the upper lag limit h_2 using Theorem 1 [5], it is assumed that the lower lag limit h_1 is divided into m number of partitions such that $h_1 = \tau m$ where integer τ is partition size. From Table 1, it is also clear that the Theorem 1 provides less conservative results than the previous results [4-6].

Table 1. Upper lag limit h_2 for different lower lag limit h_1 for the system taken in Example 1.

Methods/ h_1	2	6	8	12
Theorem 1 [4]	8	8	10	13
Theorem 1 [5]	8 ($m=2, \tau=1$)	9 ($m=3, \tau=2$)	10 ($m=4, \tau=2$)	13 ($m=6, \tau=2$)
Theorem 1 [6]	9	10	11	14
Theorem 1	9	11	12	15

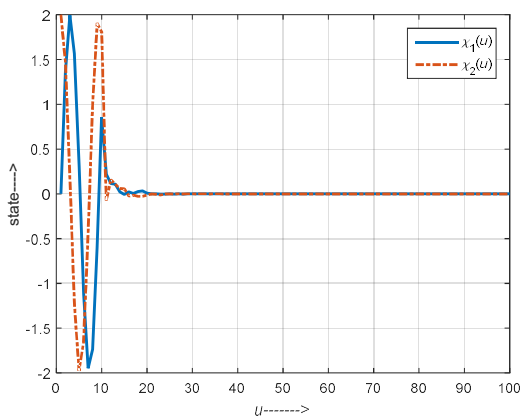


Fig. 1. State trajectories of the system considered in Example 1.

The state trajectories of the chosen system are depicted in Figure 1 where $\mathbf{F}_0 = \mathbf{F}_1 = \mathbf{I}$, $2 \leq d(u) \leq 9$ and initial condition is selected randomly. Moreover, in Figure 1, $\chi(u) \rightarrow \mathbf{0}$ as $u \rightarrow \infty$ implying the global asymptotic stability of the considered system.

Example 2. Choose the system (63) together with

$$\mathbf{A} = \begin{bmatrix} 0.8 & 0 \\ 0 & 0.91 \end{bmatrix}, \mathbf{A}_d = \begin{bmatrix} -0.1 & 0 \\ -0.1 & -0.1 \end{bmatrix} \quad (69a)$$

$$\mathbf{H}_0 = \mathbf{H}_1 = \begin{bmatrix} 0 \\ 0.1 \end{bmatrix}, \mathbf{E}_0 = [0.01 \ 0], \mathbf{E}_1 = [0 \ 0.01] \quad (69b)$$

and $[k_o, k_q] = [0 \ 1]$. Using Corollary 1, it is verified that the system under consideration is GAS over the delay $h = 18$.

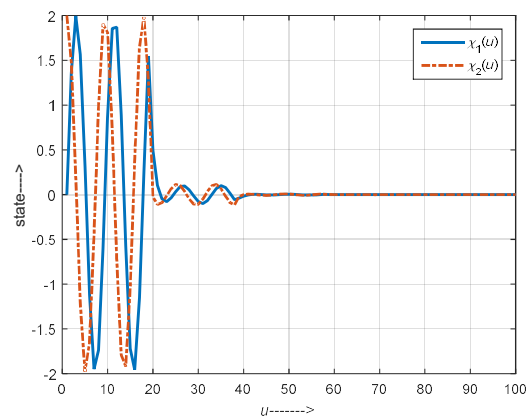


Fig. 2. State trajectories of the system considered in Example 2.

In Figure 2, the state trajectories of the present system are depicted where initial condition is assumed arbitrarily and $\mathbf{F}_0 = \mathbf{F}_1 = \mathbf{I}$. Further, in Figure 2, $\chi(u) \rightarrow \mathbf{0}$ as $u \rightarrow \infty$ which implies the global asymptotic stability of the present system.

5 Conclusion

This paper establishes delay-dependent LMI-based stability conditions for the discrete-time uncertain systems having combination of quantization and overflow nonlinearities and variable time-lags. The approach given in this paper, relative to previous approaches, helps in reducing conservativeness which in turn provides a larger stability region. Numerical examples are also given for proving the efficacy of the proposed conditions.

6 Future Scope

The approach presented in this paper can be extended to derive stability criteria for a category of discrete-time systems involving external disturbances, in addition to nonlinearities and variable time-lags which require further investigation. The possible utilization of the present idea for studying the stability analysis of uncertain two-dimensional systems with nonlinearities and variable time-lags appears to be more practical and challenging problem for future work.

Conflict of Interest: NIL

REFERENCES

- [1] Sim, P.K. and Pang, K.K. (1985). Design criterion for zero-input asymptotic overflow-stability of recursive digital filters in the presence of quantization. *Circuits, Systems and Signal Processing*, 4(4), 485-502.
- [2] Kar, H. and Singh, V. (2001). Stability analysis of 1-D and 2-D fixed-point state-space digital filters using any combination of overflow and quantization nonlinearities. *IEEE Transactions on Signal Processing*, 49(5), 1097-1105.
- [3] Kandanvli, V.K.R. and Kar, H. (2009). An LMI condition for robust stability of discrete-time state-delayed systems using quantization/overflow nonlinearities. *Signal Processing*, 89(11), 2092-2102.
- [4] Kandanvli, V.K.R. and Kar, H. (2011). Delay-dependent LMI condition for global asymptotic stability of discrete-time uncertain state-delayed systems using quantization/overflow nonlinearities. *International Journal of Robust and Nonlinear Control*, 21(14), 1611-1622.
- [5] Tadepalli, S.K. and Kandanvli, V.K.R. (2016). Improved stability results for uncertain discrete-time state-delayed systems in the presence of nonlinearities. *Transactions of the Institute of Measurement and Control*, 38(1), 33-43.
- [6] Tadepalli, S.K., Kandanvli, V.K.R. and Vishwakarma, A. (2018). Criteria for stability of uncertain discrete-time systems with time-varying delays and finite wordlength nonlinearities. *Transactions of the Institute of Measurement and Control*, 40(9), 2868-2880.
- [7] Bakule, L., Rodellar, J. and Rossell, J.M. (2006). Robust overlapping guaranteed cost control of uncertain state-delay discrete-time systems. *IEEE Transactions on Automatic Control*, 51(12), 1943-1950.
- [8] Chen, W.H., Guan, Z.H. and Lu, X. (2003). Delay-dependent guaranteed cost control for uncertain discrete-time systems with delay. *IEE Proceedings-Control Theory and Applications*, 150(4), 412-416.
- [9] Guan, X., Lin, Z. and Duan, G. (1999). Robust guaranteed cost control for discrete-time uncertain systems with delay. *IEE Proceedings- Control Theory and Applications*, 146(6), 598-602.
- [10] Kandanvli, V.K.R. and Kar, H. (2008). Robust stability of discrete-time state-delayed systems employing generalized overflow nonlinearities. *Nonlinear Analysis: Theory, Methods & Applications*, 69(9), 780-787.
- [11] Lu, G. and Ho, D.W.C. (2004). Robust H_∞ observer for nonlinear discrete systems with time delay and parameter uncertainties. *IEE Proceedings-Control Theory and Applications*, 151(4), 439-444.
- [12] Palhares, R. M., de Souza, C. E., and Peres, P. D. (2001). Robust H_∞ filtering for uncertain discrete-time state-delayed systems. *IEEE Transactions on Signal Processing*, 49(8), 1696-1703.
- [13] Wei, C., Chang, X. and Wang, B. (2008). Delay-dependent robust control design for uncertain multi-time-delayed linear systems. *International Journal of Intelligent Computing and Cybernetics*, 1(1), 128-142.
- [14] Wu, L., Lam, J., Yao, X. and Xiong, J. (2011). Robust guaranteed cost control of discrete-time networked control systems. *Optimal Control Applications and Methods*, 32(1), 95-112.
- [15] Xu, S. and Chen, T. (2004). Robust H_∞ control for uncertain discrete-time systems with time-varying delays via exponential output feedback controllers. *Systems & Control Letters*, 51(3-4), 171-183.
- [16] Xu, S.Y. (2002). Robust H_∞ filtering for a class of discrete-time uncertain nonlinear systems with state delay. *IEEE Transactions on Circuits and Systems Part I: Fundamental Theory and Applications*. 49(12), pp.1853.
- [17] Gupta, P. K. and Kandanvli, V. K. R. (2020). New LMI criteria to the global asymptotic stability of uncertain discrete-time systems with time delay and generalized overflow nonlinearities. In *Advances in VLSI, Communication, and Signal Processing* (pp. 883-895). Springer, Singapore.
- [18] Gupta, P. K., Singh, K. and Kandanvli, V. K. R. (2022). Further results on delay-dependent stability analysis of uncertain discrete-time systems exerting generalized overflow nonlinearities and time-varying delays. In *Recent Trends in Electronics and Communication* (pp. 1081-1100). Springer, Singapore.
- [19] Hao, F. and Zhao, X. (2010). New delay-dependent stability conditions for discrete-time systems with time-varying delay in the state. *IMA Journal of Mathematical Control and Information*, 27(3), 253-266.
- [20] He, Y., Wu, M., Liu, G.P. and She, J.H. (2008). Output feedback stabilization for a discrete-time system with a time-varying delay. *IEEE Transactions on Automatic Control*, 53(10), 2372-2377.
- [21] Jiang, X., Han, Q. L., and Yu, X. (2005). Stability criteria for linear discrete-time systems with interval-like time-varying delay. In *Proceedings of the 2005, American Control Conference, 2005*. (pp. 2817-2822). IEEE.
- [22] Li, X., Zhang, X. and Wang, X. (2017). Stability analysis for discrete-time Markovian jump systems with time-varying delay: A homogeneous polynomial approach. *IEEE Access*, 5, 27573-27581.
- [23] Nam, P.T., Pathirana, P.N. and Trinh, H. (2015). Discrete Wirtinger-based inequality and its application. *Journal of the Franklin Institute*, 352(5), 1893-1905.
- [24] Park, P., Ko, J.W. and Jeong, C. (2011). Reciprocally convex approach to stability of systems with time-varying delays. *Automatica*, 47(1), 235-238.
- [25] Seuret, A., Gouaisbaut, F. and Fridman, E. (2015). Stability of discrete-time systems with time-varying delays via a novel summation inequality. *IEEE Transactions on Automatic Control*, 60(10), 2740-2745.
- [26] Wu, M., Peng, C., Zhang, J., Fei, M. and Tian, Y. (2017). Further results on delay-dependent stability criteria of discrete systems with an interval time-varying delay. *Journal of the Franklin Institute*, 354(12), 4955-4965.

- [27] Zhang, C.K., He, Y., Jiang, L., Wang, Q.G. and Wu, M. (2017). Stability analysis of discrete-time neural networks with time-varying delay via an extended reciprocally convex matrix inequality. *IEEE Transactions on Cybernetics*, 47(10), 3040-3049.
- [28] Zhang, C.K., He, Y., Jiang, L., Wu, M. and Zeng, H.B. (2017). Summation inequalities to bounded real lemmas of discrete-time systems with time-varying delay. *IEEE Transactions on Automatic Control*, 62(5), 2582-2588.
- [29] Zhang, R., Li, J. and Jiao, J. (2016). Absolute stability of Lurie systems with two additive time-varying delays. *IMA Journal of Mathematical Control and Information*, 35(2), 555-567.
- [30] Zong, G. and Hou, L. (2010). New delay-dependent stability result and its application to robust performance analysis for discrete-time systems with delay. *IMA Journal of Mathematical Control and Information*, 27(3), 373-386.
- [31] Gupta, P.K., Singh, K., Kandanvli, V.K.R. and Kar, H. (2023). New criterion for the stability of discrete-time systems with state saturation and time-varying delay. *Journal of Control Automation and Electrical System*, 34(4), 700-708.
- [32] Singh, K., Gupta, P.K., Chaurasia, D. and Kandanvli, V. K. R. (2022). Stability of discrete-time delayed systems subject to external interference and generalized overflow nonlinearities. *IEEE Transactions on Industry Applications*, 58(4), 5353-5364.
- [33] Yue, D., Tian, E., Wang, Z. and Lam, J. (2009). Stabilization of systems with probabilistic interval input delays and its applications to networked control systems. *IEEE Transactions on Systems, Man, and Cybernetics-Part A: Systems and Humans*, 39(4), 939-945.
- [34] Seuret, A. and Gouaisbaut, F. (2013). Wirtinger-based integral inequality: application to time-delay systems. *Automatica*, 49(9), 2860-2866.
- [35] Zhang, X.M. and Han, Q. L. (2015). Abel lemma-based finite-sum inequality and its application to stability analysis for linear discrete time-delay systems. *Automatica*, 57, 199-202.
- [36] Zhang, D., Shi, P., Zhang, W.A. and Yu, L. (2017). Energy-efficient distributed filtering in sensor networks: a unified switched system approach. *IEEE Transactions on Cybernetics*, 47(7), 1618-1629.
- [37] Boyd, S., Ghaoui, L.E., Feron, E. and Balakrishnan, V. (1994). *Linear matrix inequalities in system and control theory*. Philadelphia, PA: SIAM
- [38] Xu, S., Lam, J., Lin, Z. and Galkowski, K. (2002). Positive real control for uncertain two-dimensional systems. *IEEE Transactions on Circuits and Systems I: Fundamental Theory and Applications*, 49(11), 1659-1666.
- [39] Löfberg, J. (2004). YALMIP: A toolbox for modeling and optimization in MATLAB. In *2004 IEEE international conference on robotics and automation*, Taipei, Taiwan, (pp. 284-289).